

Features

- Full 16-Bit Performance
- 2.7-V – 5.5-V Single-Supply Operation
- High Accuracy: INL 1LSB
- Fast Settling Speed: 1 μ s
- 10-nV/ $\sqrt{\text{Hz}}$ Output Noise Density
- Unbuffered Voltage Output
- SPI Compatible Interface
- Power-On Reset to 0 V (TPC2160 and TPC2161) or mid-scale (TPC2161M)
- Low Glitch: 10 nV-sec
- TPC2160 Package: SOP-8
- TPC2161/TPC2161M Package: SOP-14

Applications

- Data Acquisition Systems
- Automatic Test Equipment
- Industrial Process Control

Description

The TPC2160 and TPC2161/TPC2161M are single-channel, 16-bit, voltage output digital-to-analog converters with SPI interface. They accept a wide supply voltage range.

The TPC2160 output is 0 V to V_{REF} , and the TPC2161/TPC2161M can provide bipolar output $\pm V_{\text{REF}}$ with external buffer and internal feedback resistor ladder.

These parts incorporate a power-on reset circuit to ensure that the DAC output powers up to 0 V (TPC2160 and TPC2161), or mid-scale (TPC2161M).

The DACs provide 16-bit resolution over the full specified temperature range of -40°C to 105°C .

The DACs achieve a 1- μ s settling time. The outputs are unbuffered, with low power consumption and low offset errors.

Providing a low noise performance of 10 nV/ $\sqrt{\text{Hz}}$ and low glitch, the DACs are suitable for deployment across multiple end systems.

Functional Block Diagrams

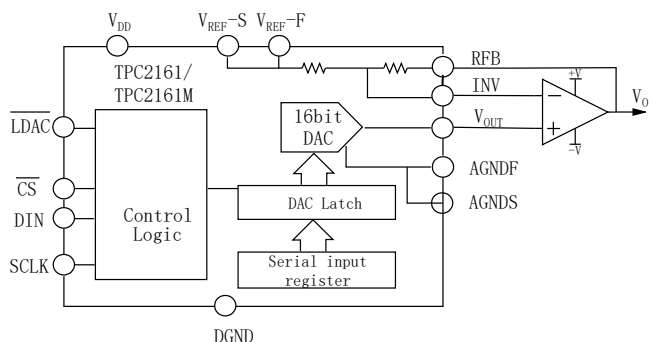
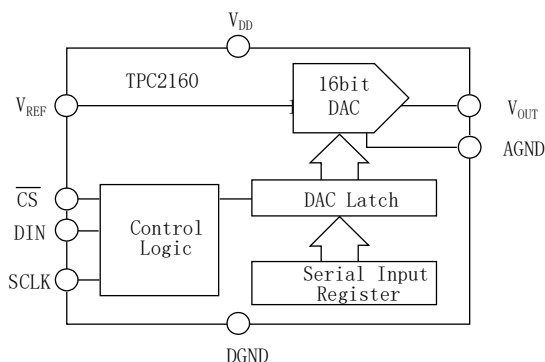


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Product Family Table

Order Number	Resolution	Reference	Output POR status	Package
TPC2160	16	External	0	SOP8, DFN3X3-8
TPC2161	16	External	0	SOP14
TPC2161M	16	External	Midscale	SOP14

Revision History

Date	Revision	Notes
2021-06-16	Rev.A.1	Initial version
2021-08-24	Rev.A.2	Updated Absolute Maximum Ratings
2022-01-17	Rev.A.3	Updated reference input impedance
2022-12-27	Rev.A.4	Corrected TPC2161 name suffix
2023-04-03	Rev.A.5	Updated VIH/VIL threshold
2024-02-21	Rev.A.6	Added TPC2161M information
2025-11-03	Rev.A.7	- Added the DFN3X3-8 Package and Application Information - Updated Timing Characteristics and Timing Diagrams

Pin Configuration and Functions

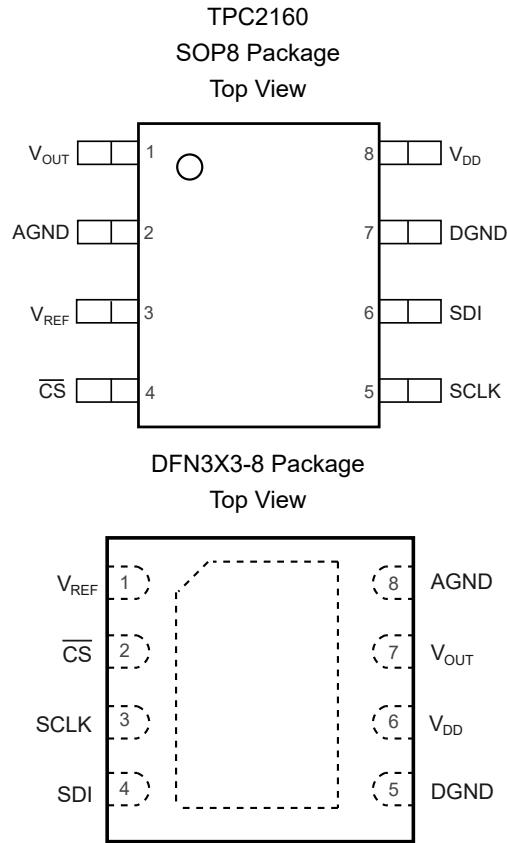


Table 1. Pin Functions: TPC2160

Pin No.		Pin Name	Description
SOP8	DFN3X3-8		
1	7	V _{OUT}	DAC analog output
2	8	AGND	Analog Ground
3	1	V _{REF}	Voltage Reference Input for the DAC
4	2	$\overline{\text{CS}}$	Chip select input (active low)
5	3	SCLK	Clock Input. Data is clocked into the input register on the rising edge of SCLK
6	4	SDI	Serial Data Input
7	5	DGND	Digital Ground
8	6	V _{DD}	Analog Supply Voltage

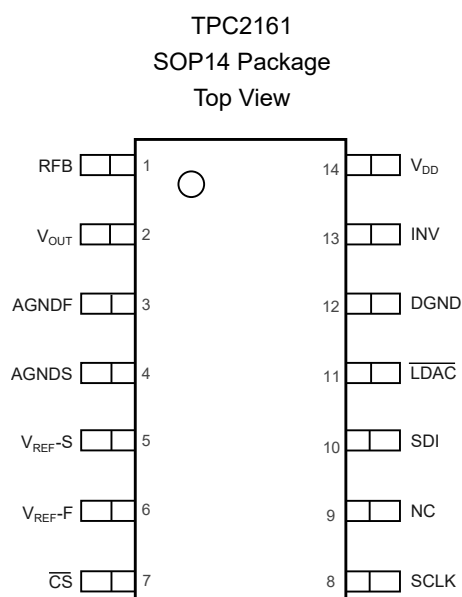


Table 2. Pin Functions: TPC2161

Pin No.	Pin Name	Description
1	RFB	Feedback resistor. Connect to the output of the external operational amplifier in bipolar mode.
2	V _{OUT}	Analog output of DAC
3	AGNDF	Analog ground (Force)
4	AGNDS	Analog ground (Sense)
5	V _{REF-S}	Voltage reference input (Sense). Connect to an external voltage reference.
6	V _{REF-F}	Voltage reference input (Force). Connect to an external voltage reference.
7	$\overline{CS}$	Chip select input (active low). Data is not clocked into SDI unless $\overline{CS}$ is low.
8	SCLK	Serial clock input
9	NC	No internal connection
10	SDI	Serial data input. Data are latched into the input register on the rising edge of SCLK
11	$\overline{LDAC}$	Load DAC control input. Active low. When $\overline{LDAC}$ is Low, the DAC latch is simultaneously updated with the content of the input register.
12	DGND	Digital ground
13	INV	Junction point of internal scaling resistors. Connect to an external operational amplifier inverting input in bipolar mode.
14	V _{DD}	Analog power supply, +3 V to +5 V

Specifications

Absolute Maximum Ratings ⁽¹⁾

Parameter		Min	Typ	Max	Unit
Supply Voltage	$V^+ - V^-$	-0.3		6	V
	Analog Input Voltage	-0.3		$V^+ + 0.3$	V
	Digital Input Voltage to DGND	-0.3		$V^+ + 0.3$	V
	V_{OUT} to AGND	-0.3		$V^+ + 0.3$	V
	AGND, AGNDF, AGNDS to DGND	-0.3		+0.3	V
	Input Current: +IN, -IN ⁽²⁾	-10		+10	mA
	Output Current: OUT	-10		+10	mA
	Operating Temperature Range	-40		125	°C
	Maximum Junction Temperature			150	°C
	Storage Temperature Range	-65		150	°C

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

(2) The inputs are protected by ESD protection diodes to each power supply.

ESD, Electrostatic Discharge Protection

Symbol	Parameter	Condition	Minimum Level	Unit
HBM	Human Body Model ESD	ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	5.5	kV
CDM	Charged Device Model ESD	ANSI/ESDA/JEDEC JS-002 ⁽²⁾	1.5	kV

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

Recommended Operating Conditions

		MIN	NOM	MAX	UNIT
V_{DD}	Analog Supply Voltage	2.7		5.5	V
V_{IO}	Digital IO Supply Voltage	1.7		5.5	V
Digital input voltage	Digital Input Voltage	0		V_{IO}	V
VREFIN	Reference Divider Disabled	1.2		$(V_{DD}-0.2)/2$	V
	Reference Divider Enabled	2.4		$V_{DD}-0.2$	V
	Reference Divider Disabled	1.2		$V_{DD}/2$	V
	Reference Divider Enabled	2.4		V_{DD}	V
T_J	Operating Junction Temperature	-40		125	°C

Thermal Information

Package Type	θ_{JA}	θ_{JC}	Unit
SOP8	112.4	64.1	°C/W
SOP14	96.7	46.7	°C/W
DFN3X3-8	60	49	°C/W

Electrical Characteristics

All test conditions: $V_{DD} = 2.7\text{ V to }5.5\text{ V}$, $2.5\text{ V} \leq V_{REF} \leq V_{DD}$, $AGND = DGND = 0\text{ V}$. All specifications $T_A = -40\text{ to }105\text{ }^{\circ}\text{C}$, unless otherwise noted.

Parameter	Test conditions	Min	Typ	Max	Unit
Static Performance					
Linearity Error				± 1	LSB
Differential Linearity Error				± 1	LSB
Gain Error			± 0.5	± 7	LSB
Gain Drift			± 0.1		ppm/ $^{\circ}\text{C}$
Zero Code Error			± 0.3	± 2	LSB
Zero Code Drift			± 0.05		ppm/ $^{\circ}\text{C}$
Output Characteristics					
Voltage Output		0		V_{DD}	V
Output Impedance			6.25		k Ω
Feedback Resistor (TPC2161)	RFB, RINV		28		k Ω
Bipolar Resistor Matching	RFB / RINV		1		Ω/Ω
	Ratio error		0.01		%
Dynamic Performance					
Settling Time	To 1/2 LSB of FS, $CL = 10\text{ pF}$		1		μs
Slew Rate	$CL = 10\text{ pF}@5\text{ V}$		25		V/ μs
Digital-To-Analog Glitch	1 LSB change around major carry		10		nV-s
Digital Feedthrough			0.2		nV-s
Output Noise (TPC2160)	DAC code = 0x8400, frequency = 1 kHz $T_A = +25^{\circ}\text{C}$		10		nV/ $\sqrt{\text{Hz}}$
Output Noise (TPC2161)	DAC code = 0x8400, frequency = 1 kHz $T_A = +25^{\circ}\text{C}$		18		nV/ $\sqrt{\text{Hz}}$
Power-Supply Rejection	V_{DD} varies $\pm 10\%$		± 0.1		LSB
Reference Input ⁽²⁾					
Reference Input Voltage Range		1.25		V_{DD}	V
Reference Input Impedance ⁽¹⁾ (TPC2160)	Unipolar mode	8.5			k Ω
Reference Input Impedance ⁽¹⁾ (TPC2161)	Unipolar mode	6.5			k Ω
Reference -3dB Bandwidth, BW	Code = FFFFh		1.3		MHz
Reference Feedthrough	Code = 0000h, $V_{REF} = 1\text{ VPP}$ at 100 kHz		1		mV

Parameter	Test conditions	Min	Typ	Max	Unit
Reference Input ⁽²⁾					
Signal-to-Noise Ratio, SNR			92		dB
Reference Input Capacitance	Code = 0000h		75		pF
	Code = FFFFh		120		pF
Digital Inputs					
VIL Input Low Voltage	V _{DD} = 2.7 V			0.6	V
	V _{DD} = 5 V			0.8	
VIH Input High Voltage	V _{DD} = 2.7 V	2.1			V
	V _{DD} = 5 V	2.4			
Input Current			±0.5	5	μA
Input Capacitance ⁽²⁾				10	pF
Hysteresis Voltage ⁽²⁾			0.4		V
Power Supply					
V _{DD} Power-Supply Voltage		2.7		5.5	V
I _{DD} Power-Supply Current	V _{DD} = 5			150	μA
Power	V _{DD} = 5			825	μW
SPI					
Fclk				50	MHz

(1) Reference input resistance is code-dependent, minimum at 0x8555.

(2) Guaranteed by design, not subject to production test.

Timing Characteristics

All test conditions are: $V_{DD} = 2.7\text{ V to }5.5\text{ V} \pm 10\%$, $V_{REF} = 2.5\text{ V}$, $V_{INH} = 3\text{ V}$ and 90% of V_{DD} , $V_{INL} = 0\text{ V}$ and 10% of V_{DD} , $AGND = DGND = 0\text{ V}$; $T_A = -40\text{ to }105\text{ }^{\circ}\text{C}$, unless otherwise noted.

Parameter ⁽¹⁾ (2)	Limit	Unit	Description
f_{SCLK}	50	MHz max	SCLK cycle frequency
t_{SCK}	20	ns min	SCLK cycle time
t_{WSCK}	10	ns min	SCLK high or low time
t_{td}	30	ns min	$\overline{CS}$ high during active period
t_{Lead}	10	ns min	$\overline{CS}$ low to SCLK high setup
t_{DSCLK}	15	ns min	$\overline{CS}$ high to SCLK high setup
t_{Delay}	30	ns min	SCLK high to $\overline{CS}$ low hold time
t_{Lag}	20	ns min	SCLK high to $\overline{CS}$ high hold time
t_{su}	15	ns min	Data setup time
t_{ho}	4	ns min	Data hold time ($V_{INH} = 90\%$ of V_{DD} , $V_{INL} = 10\%$ of V_{DD})
	7.5	ns min	Data hold time ($V_{INH} = 3\text{ V}$, $V_{INL} = 0\text{ V}$)
t_{WLDAC}	30	ns min	$\overline{LDAC}$ pulse width
t_{DLDAC}	30	ns min	$\overline{CS}$ high to $\overline{LDAC}$ low setup

(1) Parameters are provided by design simulation.

(2) All input signals are specified with $t_R = t_F = 1\text{ ns/V}$ and timed from a voltage level of $(V_{INL} + V_{INH})/2$.

Timing Diagrams

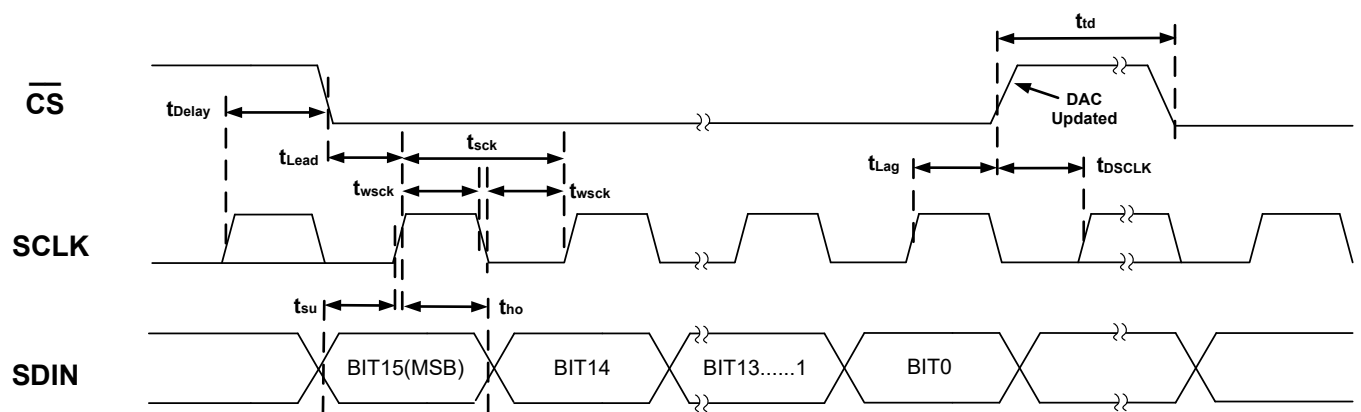
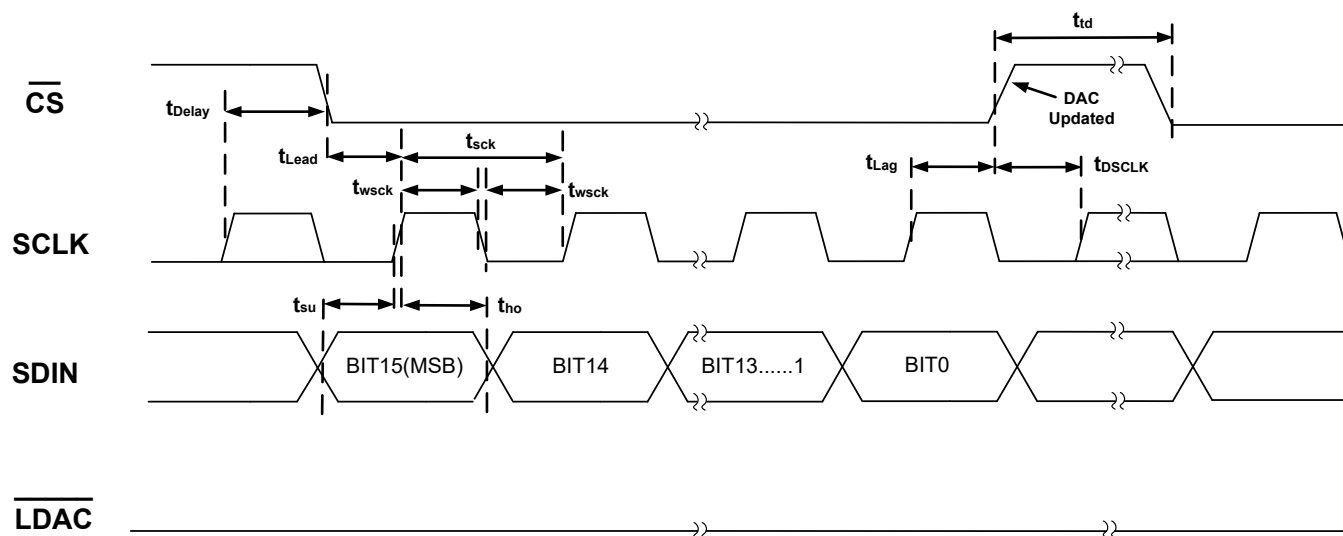


Figure 1. TPC2160 SPI Timing Diagram

case1: $\overline{LDAC}$ tied to LOW



case2: $\overline{LDAC}$ Active

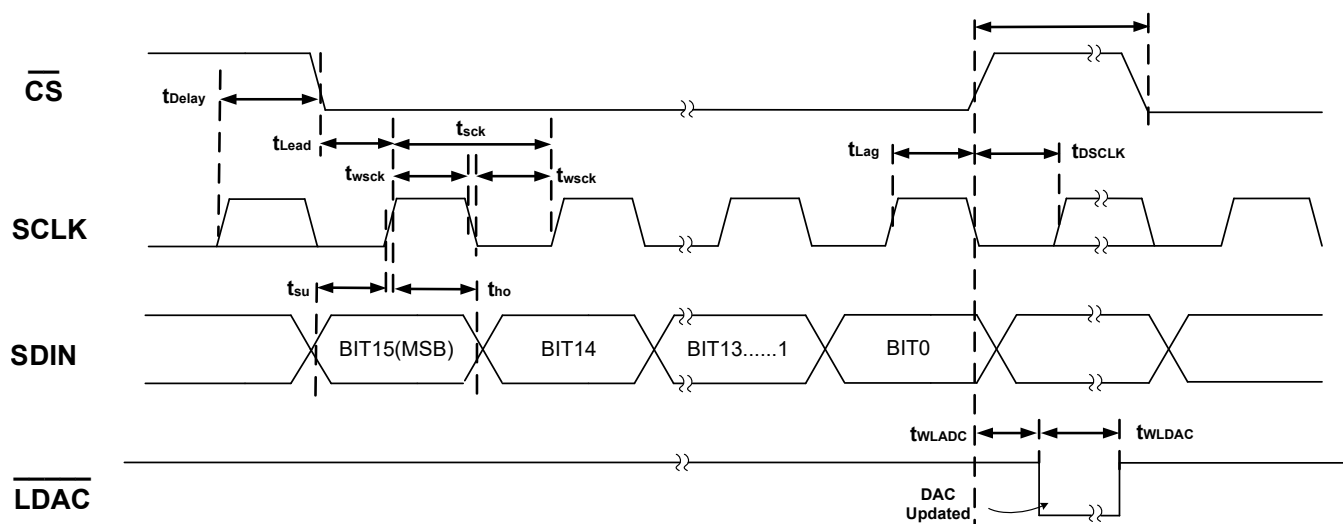


Figure 2. TPC2161 SPI Timing Diagram

Typical Performance Characteristics

All test conditions: $T_A = 25^\circ\text{C}$, $V_{DD} = 5.5\text{ V}$, Internal Reference = 2.5 V, Gain = 2, DAC outputs unloaded, unless otherwise noted.

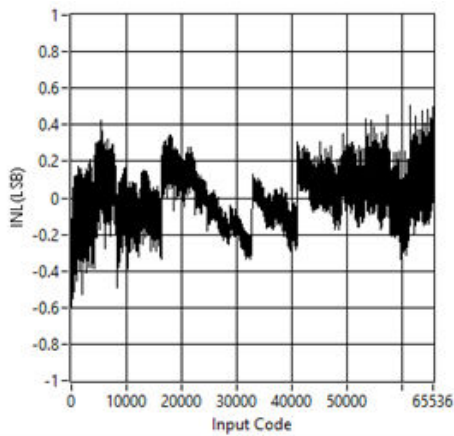


Figure 3. INL

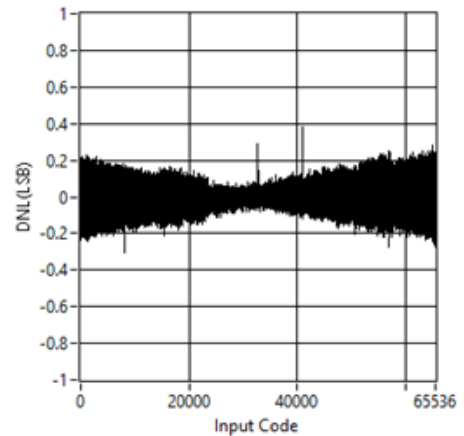


Figure 4. DNL

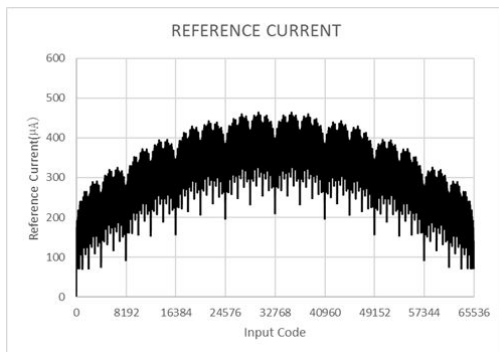


Figure 5. Reference Current

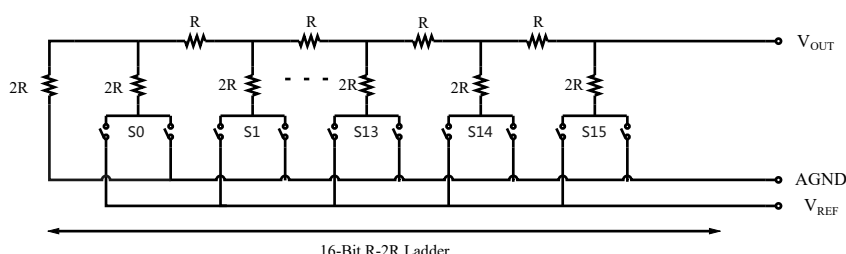
Detailed Description

Overview

The TPC2160 and TPC2161 are single-channel, 16-bit, DACs with R-2R structure. They have an SPI serial interface, with 16-bit word format. The TPC2160 and TPC2161 are reset to zero code.

Digital-to-Analog Sections

A simplified DAC diagram is shown below. The 16 bits of the data word drive switches S0 to S15 of a 16-bit voltage mode R-2R ladder network.



Feature Description

Output Range

The output of the DAC is:

$$V_{OUT} = (V_{REF} \times \text{Code}) / 65536 \quad (1)$$

Where *Code* is the decimal data word loaded to the DAC latch.

Power-On Reset

The devices have a power-on reset function, to make sure the output is a known state at power-up, and the DAC Registers are zero (TPC2160 and TPC2161) or mid-scale (TPC2161M) until new data are loaded. Therefore, after power-up, the output of V_{OUT} is 0 V (TPC2160 and TPC2161), or mid-scale (TPC2161M).

Serial Interface

The digital interface is a standard 3-wire serial interface compatible with SPI.

When $\overline{CS}$ turns low, the transmission is started, and the SDI data is shifted in and latched on the edge of SCLK. The data registers are 16-bit, so $\overline{CS}$ must go high after 16 SCLKs transfers a whole data word.

For the TPC2160, the input register is latched to the DAC immediately after the input register is loaded, so the DAC output is updated at the same time.

The TPC2161 has an $\overline{\text{LDAC}}$ pin. After $\overline{\text{CS}}$ goes high, the DAC register can be updated by bringing $\overline{\text{LDAC}}$ low. $\overline{\text{LDAC}}$ can also be tied low permanently. In this case, the DAC register is updated immediately after the input register is loaded, and the DAC output is updated at the same time.

Application and Implementation

Note

Information in the following application sections is not part of the 3PEAK's component specification and 3PEAK does not warrant its accuracy or completeness. 3PEAK's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

Application Information

Multiple Devices Decoding

The $\overline{CS}$ pin of the device can be used to select one of multiple DACs. All devices can share and receive the same serial clock and serial data, but only one device receives the $\overline{CS}$ signal at any one time.

Unipolar Output Operation

This DACs are capable of driving unbuffered loads up to 60 k Ω . When operating in the unbuffered mode, the devices deliver low supply current (typically 5 μ A) alongside low offset error. Specifically, the TPC2160 offers a unipolar output swing that spans from 0 V to V_{REF} , while the TPC2161 can be configured to output both unipolar and bipolar voltages. Typical unipolar output voltage circuits for each device are illustrated in [Figure 6](#) and [Figure 7](#), respectively. For the coding specifications corresponding to this operating mode, refer to [Table 3](#).

Table 3.

DAC LATCH CONTENTS		ANALOG OUTPUT
MSB	LSB	
1111 1111 1111 1111		$V_{REF} \times (65535/65536)$
1000 0000 0000 0000		$V_{REF} \times (32768/65536) = 1/2 V_{REF}$
0000 0000 0000 0001		$V_{REF} \times (1/65536)$
0000 0000 0000 0000		0 V

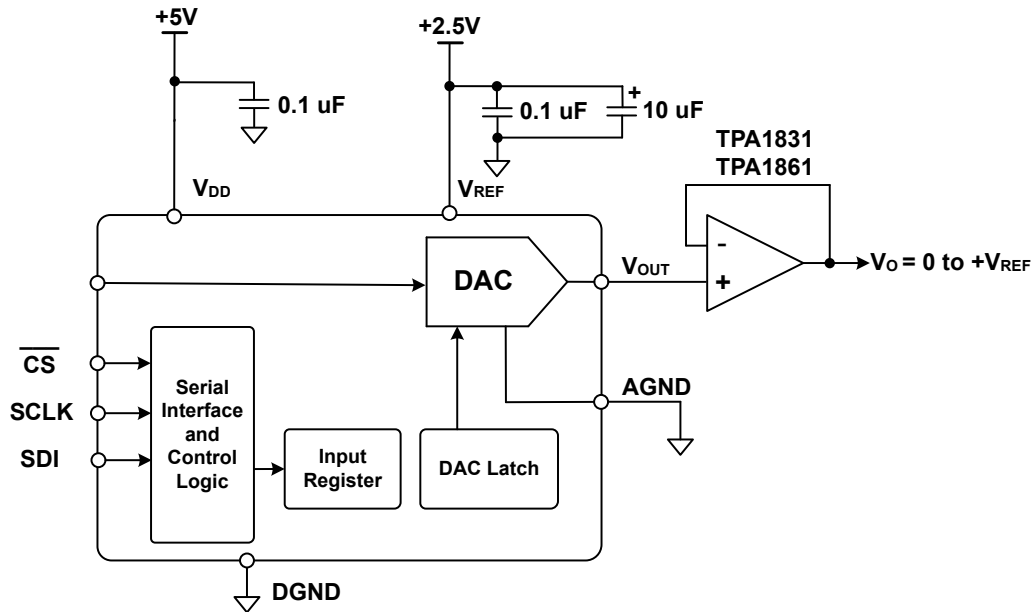


Figure 6. Unipolar Output Mode of TPC2160

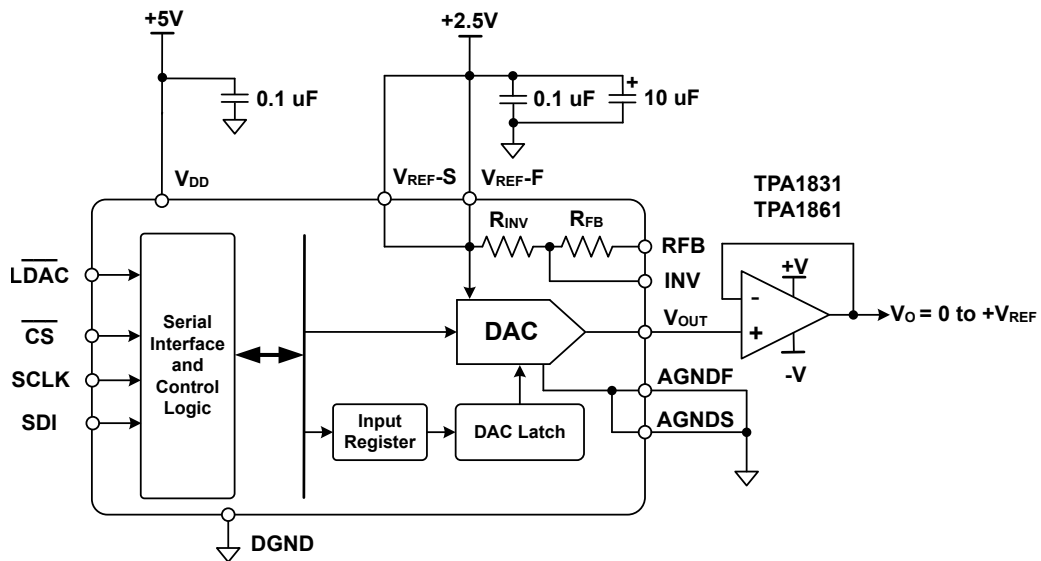


Figure 7. Unipolar Output Mode of TPC2161

Assuming a perfect reference, the worst-case output voltage may be calculated from the following equation:

$$V_{OUT_UNI} = \frac{D}{2^{16}} \times (V_{REF} + V_{GE}) + V_{ZSE} + INL \quad (2)$$

Where

- V_{OUT_UNI} = Unipolar mode worst-case output
- D = Code loaded to DAC
- V_{REF} = Reference voltage applied to part
- V_{GE} = Gain error in volts
- V_{ZSE} = Zero-scale error in volts

- INL = Integral nonlinearity in volts

Bipolar Output Operation

The TPC2161 can be configured to deliver a bipolar voltage output with the assistance of an external operational amplifier. A typical circuit for this operating mode is presented in [Figure 8](#). To enable the bipolar output swing, the matched bipolar offset resistors (R_{FB} and R_{INV}) must be connected to the external operational amplifier; typically, both R_{FB} and R_{INV} have a resistance of 28 k Ω .

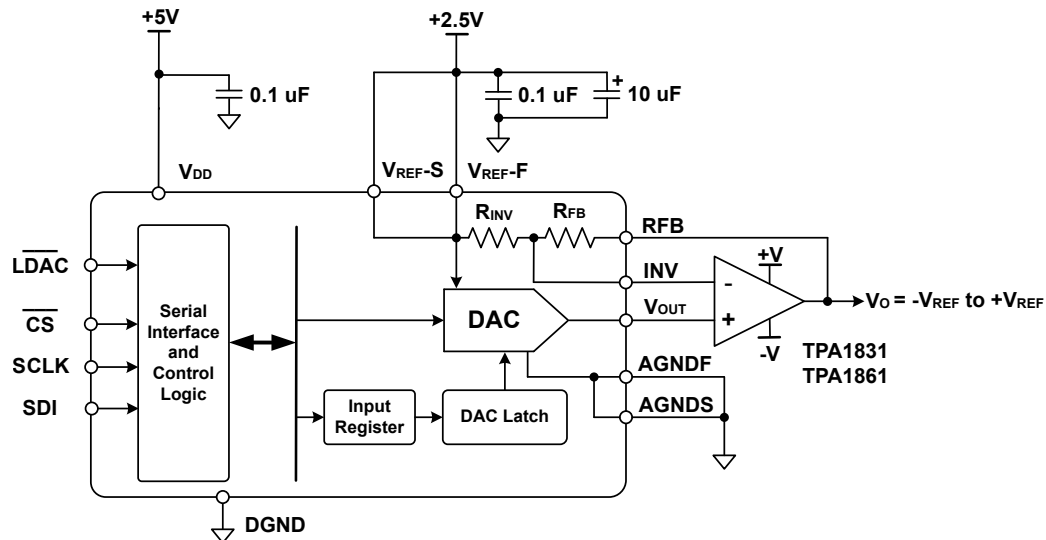


Figure 8. Bipolar Output Mode of TPC2161

Table 4 shows the transfer function for this output operating mode. The TPC2161 also provides a set of Kelvin connections to the analog ground and external reference inputs.

Table 4.

DAC LATCH CONTENTS		A/D CONVERTER OUTPUT
MSB	LSB	
1111 1111 1111 1111		+ V _{REF} X (32767/32768)
1000 0000 0000 0001		+ V _{REF} X (1/32768)
1000 0000 0000 0000		0 V
0111 1111 1111 1111		- V _{REF} X (1/32768)
0000 0000 0000 0000		- V _{REF} X (32767/32768) = - V _{REF}

Assuming a perfect reference, the worst-case output voltage may be calculated from the following equation:

$$V_{OUT_BIP} = \frac{[(V_{OUT_UNI} + V_{OS})(2 + RD) - V_{REF}(1 + RD)]}{1 + \left(\frac{2 + RD}{A}\right)} \quad (3)$$

Where

- V_{OS} = External operational amplifier input offset voltage
- $RD = R_{FB}$ and R_{IN} resistor matching error
- A = Operational amplifier open-loop gain

Output Amplifier Selection

For bipolar mode operation, a precision amplifier powered by a dual power supply is recommended, as this configuration enables an output range of $\pm V_{REF}$.

In single-supply applications, selecting an appropriate operational amplifier can be more challenging. This is because the output swing of the amplifier typically excludes the negative supply rail—referred to as analog ground (AGND) in this scenario. Unless the application avoids using codes near 0, this limited output swing may lead to a certain degradation in specified performance.

The chosen amplifier must feature low offset voltage (the DAC LSB is 38 μ V with a 2.5 V reference), which eliminates the requirement for output offset trimming. Additionally, the input bias current should be low: when multiplied by the DAC's output impedance (approximately 6.25 k Ω), this current contributes to zero-code error.

Rail-to-rail input and output capabilities are mandatory for the amplifier. To ensure fast settling, the slew rate of the operational amplifier should not impede the settling time of the DAC. While the output impedance of DAC is constant and independent of the input code, the output amplifier should have the highest possible input impedance to minimize gain errors. Furthermore, the amplifier requires a 3 dB bandwidth of 1 MHz or higher. Notably, the amplifier introduces an additional time constant to the system, which increases the output settling time. Consequently, a higher amplifier 3 dB bandwidth results in a shorter effective settling time for the combined DAC and amplifier circuit.

Reference and Ground

As the input impedance is code-dependent, the reference pin must be driven by a low-impedance source. The TPC2160 and TPC2161 are designed to operate with a voltage reference that ranges from 1.25 V to VDD. Using a reference voltage below 1.25 V will lead to a reduction in device accuracy.

The full-scale output voltage of the DAC is determined by the reference voltage. [Table 3](#) and [Table 4](#) specify the analog output voltages corresponding to specific digital codes.

For optimal performance, the TPC2161 incorporates Kelvin sense connections. If the application does not require separate force and sense lines, these lines should be tied together in close proximity to the device package. This configuration minimizes voltage drops between the package pins and the internal die.

Power Supply and Reference Bypassing

For accurate high-resolution performance, it is recommended that the reference and supply pins be bypassed with a 10 μ F tantalum capacitor in parallel with a 0.1 μ F ceramic capacitor.

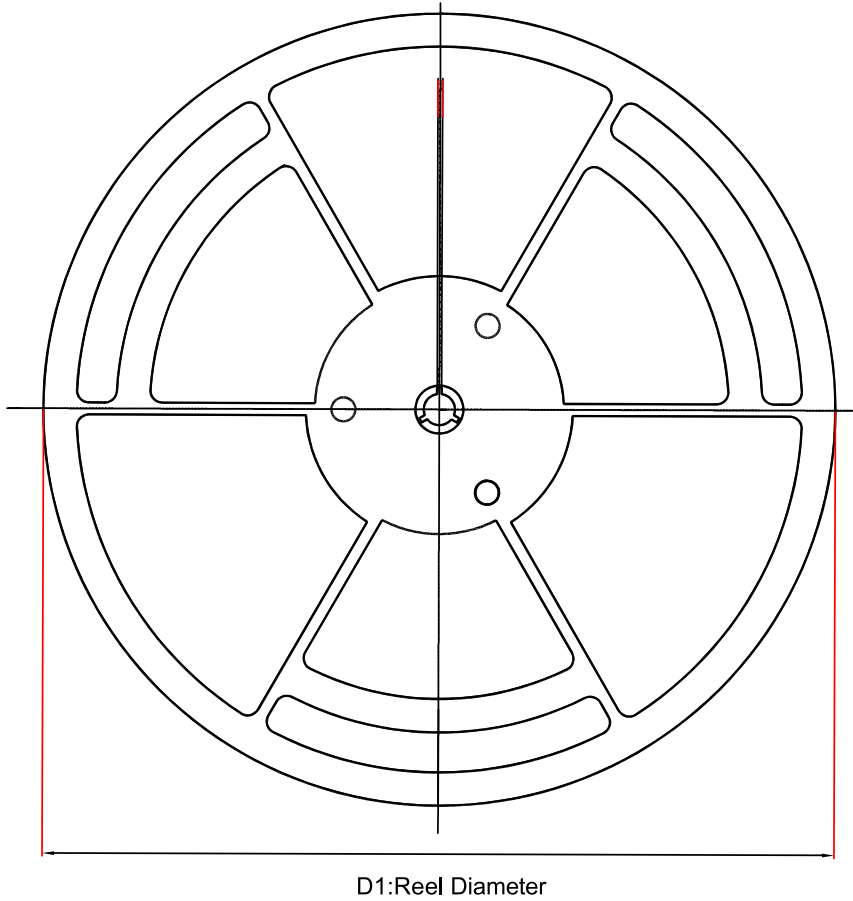
Layout

Layout Guideline

When designing the PCB for high performance circuit, the power supply and grounding layout should be carefully considered. Ensure that the analog and digital sections are clearly separated. If multiple devices require an analog ground connection to the digital ground, establish a single point of connection. Put the device close to the star ground point as close as possible.

Additionally, the device should have adequate bypass capacitors of 10 μF in parallel with 0.1 μF capacitors, placed as close to the package as feasible, ideally adjacent to the device. The 10 μF capacitors should be of the tantalum bead type, while the 0.1 μF capacitors should exhibit low effective series resistance (ESR) and inductance (ESL), similar to ceramic capacitors, providing a low-impedance path to ground at high frequencies to handle transient currents resulting from internal logic switching.

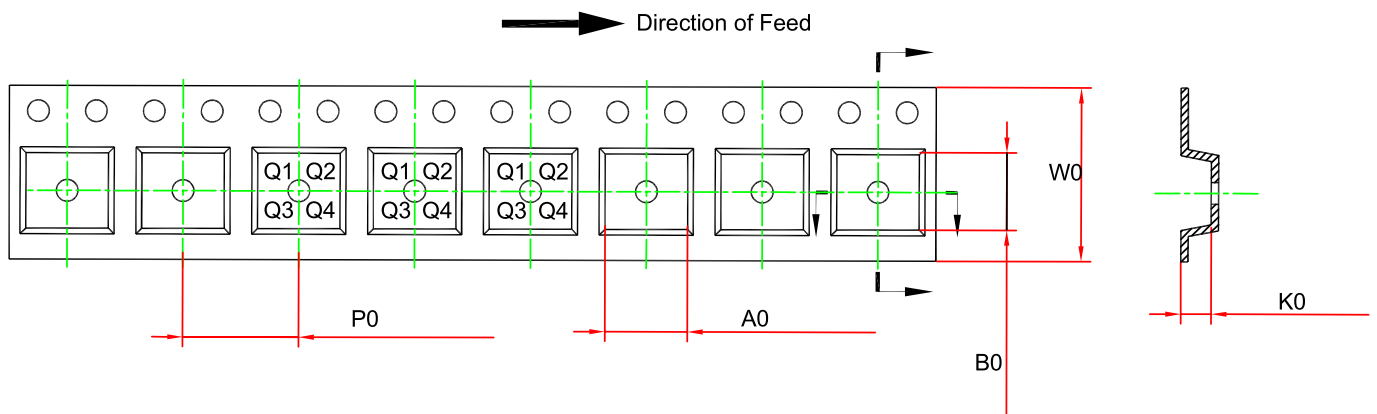
Tape and Reel Information



D1:Reel Diameter



W1

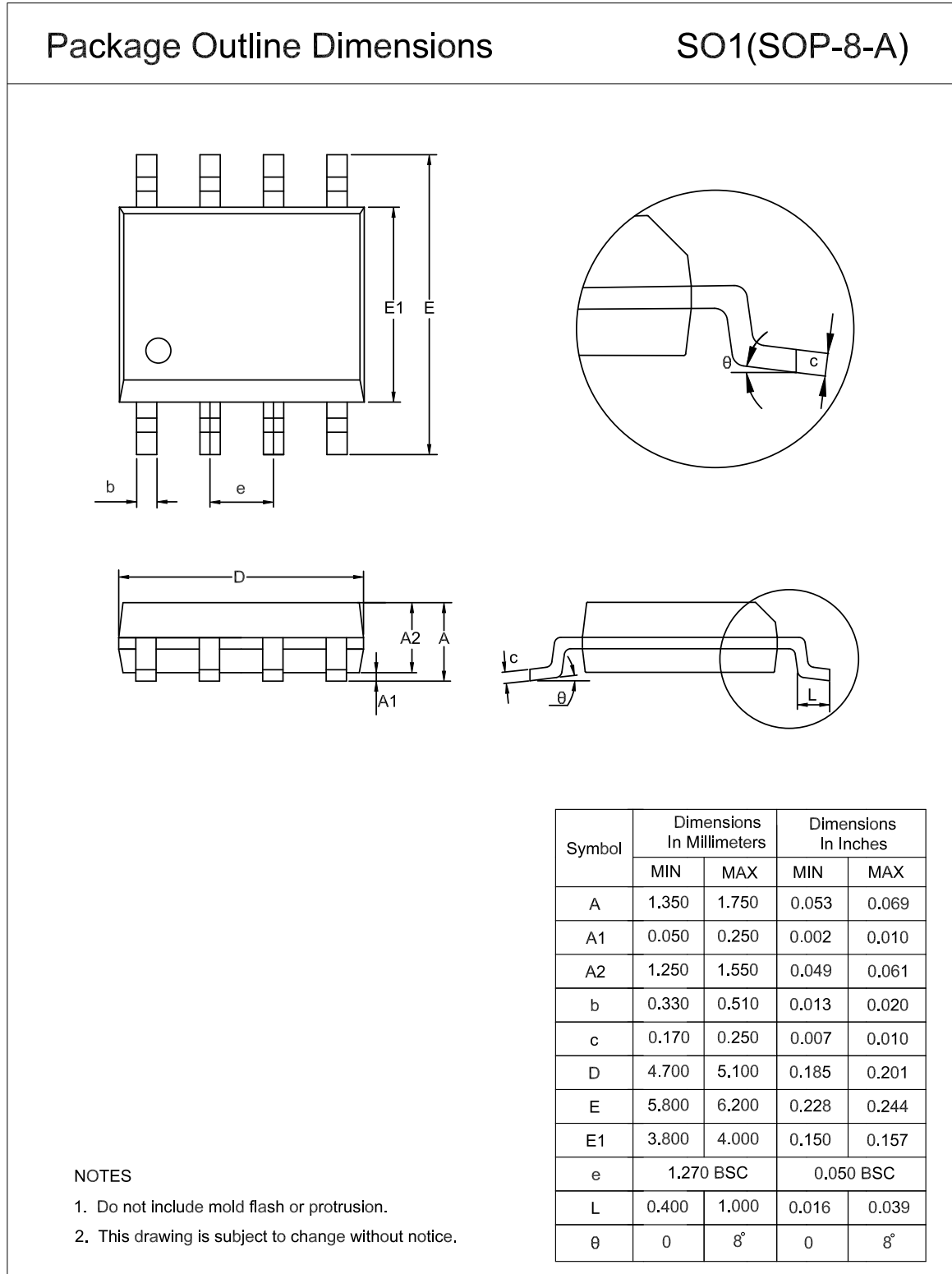


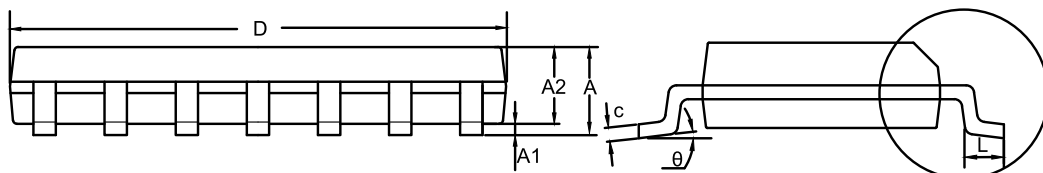
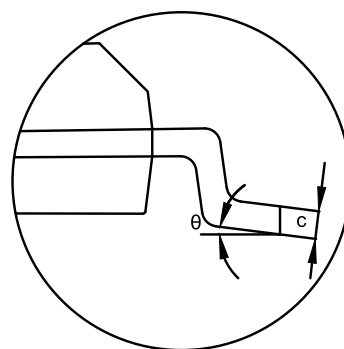
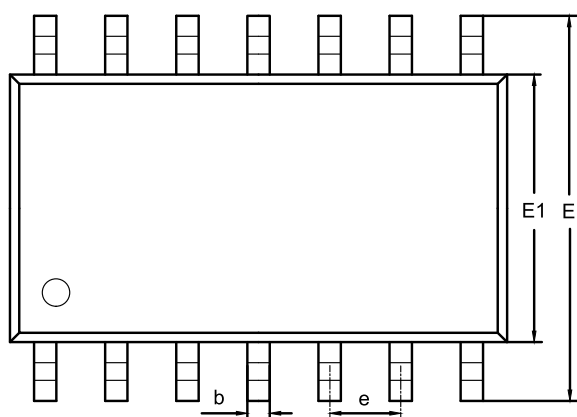
Order Number	Package	D1 (mm)	W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	W0 (mm)	Pin1 Quadrant
TPC2160S1L1A-SO1R-S	SOP8	330.0	17.6	6.4	5.4	2.1	8.0	12.0	Q1

Order Number	Package	D1 (mm)	W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	W0 (mm)	Pin1 Quadrant
TPC2160-DF6R	DFN3X3-8	330.0	17.6	3.3	3.3	1.1	8.0	12.0	Q1
TPC2161S1L1-SO2R	SOP14	330.0	21.6	6.5	9.5	2.3	8.0	16.0	Q1
TPC2161MS1L1-SO2R	SOP14	330.0	21.6	6.5	9.5	2.3	8.0	16.0	Q1

Package Outline Dimensions

SOP8



SOP14
Package Outline Dimensions
SO2(SOP-14-A)


Symbol	Dimensions In Millimeters		Dimensions In Inches	
	MIN	MAX	MIN	MAX
A	1.350	1.750	0.053	0.069
A1	0.050	0.250	0.002	0.010
A2	1.250	1.650	0.049	0.065
b	0.310	0.510	0.012	0.020
c	0.100	0.250	0.004	0.010
D	8.450	8.850	0.333	0.348
E	5.800	6.200	0.228	0.244
E1	3.800	4.000	0.150	0.157
e	1.270 BSC		0.050 BSC	
L	0.400	1.270	0.016	0.050
θ	0	8°	0	8°

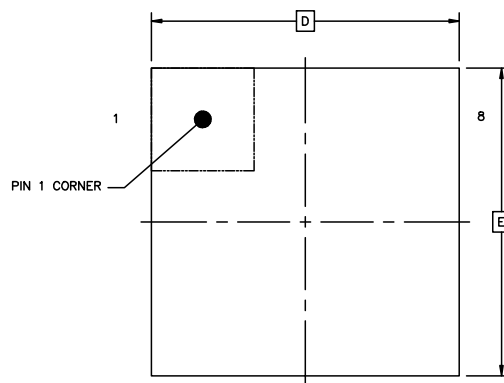
NOTES

1. Do not include mold flash or protrusion.
2. This drawing is subject to change without notice.

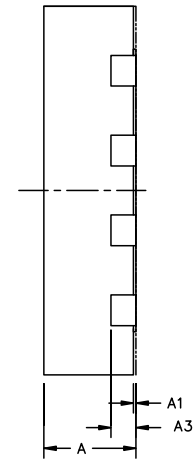
DFN3X3-8

Package Outline Dimensions

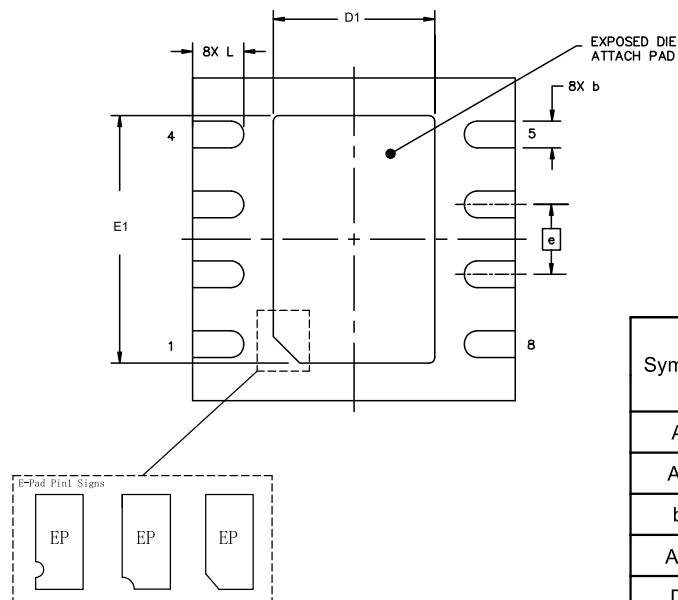
DF6(DFN3X3-8-A)



TOP VIEW



SIDE VIEW



BOTTOM VIEW

NOTES

1. Do not include mold flash or protrusion.
2. This drawing is subject to change without notice.
3. The many types of E-pad Pin1 signs may appear in the product.

Symbol	Dimensions In Millimeters		Dimensions In Inches	
	MIN	MAX	MIN	MAX
A	0.700	0.800	0.028	0.031
A1	0.000	0.050	0.000	0.002
b	0.200	0.350	0.008	0.014
A3	0.150	0.250	0.006	0.010
D	2.900	3.100	0.114	0.122
D1	1.400	1.600	0.055	0.063
E	2.900	3.100	0.114	0.122
E1	2.200	2.400	0.087	0.094
e	0.650 BSC		0.026 BSC	
L	0.224	0.575	0.009	0.023

Order Information

Order Number	Operating Temperature Range	Package	Marking Information	MSL	Transport Media, Quantity	Eco Plan
TPC2160S1L1A-SO1R-S	-40 to 125°C	SOP8	2160	1	Tape and Reel, 4000	Green
TPC2160-DF6R	-40 to 125°C	DFN3X3-8	2160	3	Tape and Reel, 4000	Green
TPC2161S1L1-SO2R	-40 to 125°C	SOP14	2161	1	Tape and Reel, 2500	Green
TPC2161MS1L1-SO2R	-40 to 125°C	SOP14	2161M	1	Tape and Reel, 2500	Green

Green: 3PEAK defines "Green" to mean RoHS compatible and free of halogen substances.

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