

**Small Size, Precision, 1.8V, 700nA Nanopower, RRIO Op-amps**  
**Description**
**Features**

- 950 nA Maximum Supply Current
- Ultra-low Single-Supply Operation Down to +1.8V
- Ground-Sensing Input Common-Mode Range
- Outputs Swing Rail-to-Rail
- Offset Voltage: 0.6 mV Maximum
- Ultra-low  $V_{OS}$  TC: 0.4  $\mu\text{V}/^\circ\text{C}$
- Ultra-low Input Bias Current: 1 pA Typical
- Stable 18 kHz GBWP with 10 mV/ $\mu\text{s}$  Slew Rate
- High 120 dB Open-Loop Voltage Gain
- Unity Gain Stable
- Outputs Source and Sink 20mA of Load Current
- No Phase Reversal for Overdriven Inputs
- Shutdown Pin Feature (TP2191N)
- $-40^\circ\text{C}$  to  $125^\circ\text{C}$  Operation Range
- Robust 8kV – HBM and 2kV – CDM ESD Rating
- Green, SC70/SOT23 Small Size Package

**Applications**

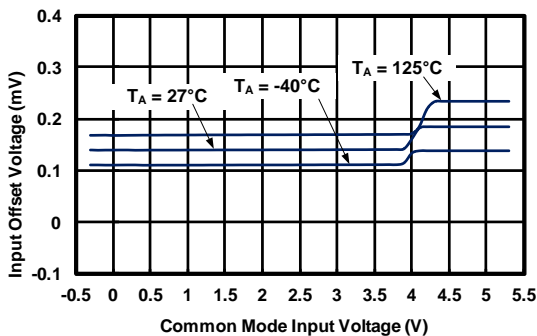
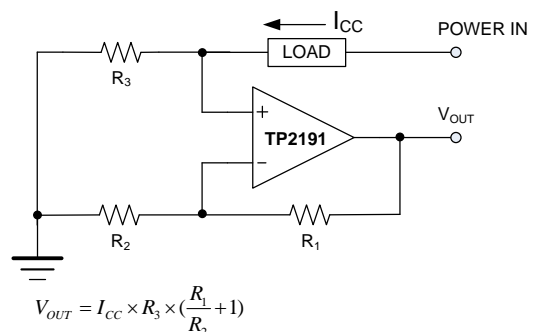
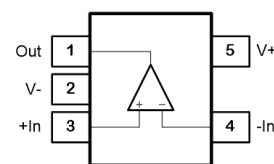
- Current Sensing
- Threshold Detectors/Discriminators
- Low Power Filters
- Wireless Remote Sensors, Active RFID Readers
- Environment/Gas/Oxygen Sensors
- Handsets and Mobile Accessories
- Battery or Solar Powered Devices
- Sensor Network Powered by Energy Scavenging

The TP2191 is a precision, ultra-low power CMOS op-amp featuring a maximum supply current of 900 nA with an ultra-low typical input bias current of 1 pA. The precision temperature compensation technique makes offset voltage temperature drift at 0.4  $\mu\text{V}/^\circ\text{C}$ , which allowing use of the TP2191 in systems with high gain without creating excessively large output offset errors.

The TP2191 is unity gain stable with a constant 18 kHz GBWP, 10 mV/ $\mu\text{s}$  slew rate, which make them appropriate for low frequency applications, such as battery current monitoring and sensor conditioning. The TP2191 can operate from a single-supply voltage of +1.8V to +6.0V or a dual-supply voltage of  $\pm 0.9\text{V}$  to  $\pm 3.0\text{V}$ . Beyond the rails input makes it very prominent in low voltage ( $< 3\text{V}$ ) rail-to-rail input applications.

The combined features make the TP2191 an ideal choice for battery-powered applications because it minimizes errors due to power supply voltage variations over the lifetime of the battery and maintain high CMRR even for a rail-to-rail input op-amp. Mobile accessories, wireless remote sensing, backup battery sensors, and single-Li+ or 2-cell NiCd/Alkaline battery powered systems can benefit from the features of the TP2191 op-amp.

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**Ground-Sensing Input Common-Mode Range**

**TP2191 in Low Side Battery Current Sensor**

# TP2191

Small Size, Precision, 1.8V, 700nA Nanopower, RRIO Op-amps

## Order Information

Order Number	Package	Transport Media, Quantity	MSL Level	Marking Information
TP2191-TR	SOT23-5	Tape and Reel, 3,000	MSL 3	B2X

## Absolute Maximum Ratings <sup>Note 1</sup>

Supply Voltage: $V^+ - V^-$ .....	6.5V	Operating Temperature Range.....	-40°C to 125°C
Input Voltage.....	$V^- - 0.3$ to $V^+ + 0.3$	Maximum Junction Temperature.....	150°C
Input Current: $+I_{IN}$ , $-I_{IN}$ <sup>Note 2</sup> .....	$\pm 10$ mA	Storage Temperature Range.....	-65°C to 150°C
Output Short-Circuit Duration <sup>Note 3</sup> .....	Indefinite	Lead Temperature (Soldering, 10 sec) .....	260°C

**Note 1:** Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

**Note 2:** The inputs are protected by ESD protection diodes to each power supply. If the input extends more than 500mV beyond the power supply, the input current should be limited to less than 10mA.

**Note 3:** A heat sink may be required to keep the junction temperature below the absolute maximum. This depends on the power supply voltage and how many amplifiers are shorted. Thermal resistance varies with the amount of PC board metal connected to the package. The specified values are for short traces connected to the leads.

## ESD, Electrostatic Discharge Protection

Symbol	Parameter	Condition	Minimum Level	Unit
HBM	Human Body Model ESD	ANSI/ESDA/JEDEC JS-001	6	kV
CDM	Charged Device Model ESD	ANSI/ESDA/JEDEC JS-002	1.5	kV

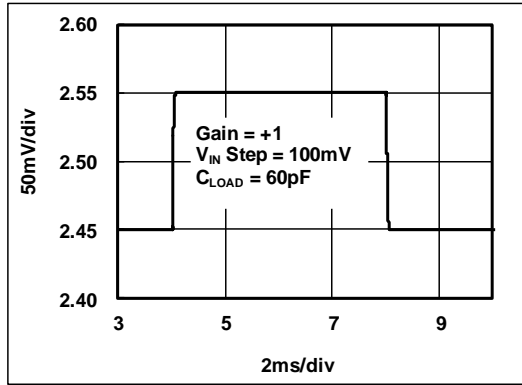
## 5V Electrical Characteristics

The specifications are at  $T_A = 27^\circ\text{C}$ .  $V_{DD} = 5\text{V}$ ,  $V_{CM} = V_{OUT} = V_{SUPPLY}/2$ ,  $R_L = 100\text{k}\Omega$ ,  $C_L = 60\text{pF}$ ,  $V_{SHDN}$  is unconnected.

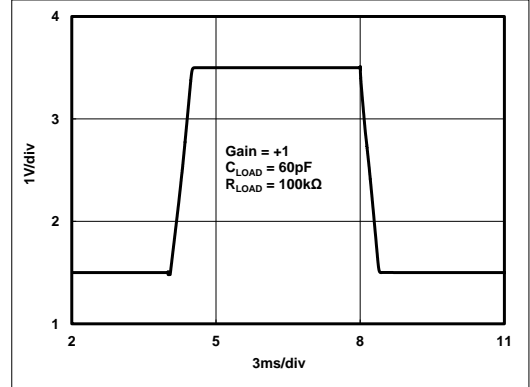
SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
$V_{OS}$	Input Offset Voltage	$V_{CM} = V_{DD}/2$	-0.6	$\pm 0.1$	+0.6	mV
$V_{OS\ TC}$	Input Offset Voltage Drift			0.4		$\mu\text{V}/^\circ\text{C}$
$I_B$	Input Bias Current	$T_A = 27^\circ\text{C}$		1		pA
		$T_A = 85^\circ\text{C}$		20		pA
		$T_A = 125^\circ\text{C}$		100		pA
$I_{OS}$	Input Offset Current			1		pA
$V_n$	Input Voltage Noise	$f = 0.1\text{Hz to } 10\text{Hz}$		6.5		$\mu\text{V}_{P-P}$
$e_n$	Input Voltage Noise Density	$f = 1\text{kHz}$		170		$\text{nV}/\sqrt{\text{Hz}}$
$R_{IN}$	Input Resistance			1		$\text{T}\Omega$
$C_{IN}$	Input Capacitance	Differential		2.9		pF
		Common Mode		5		
CMRR	Common Mode Rejection Ratio	$V_{CM} = 0.1\text{V to } 4.9\text{V}$	60	100		dB
$V_{CM}$	Common-mode Input Voltage Range		$V^- - 0.3$		$V^+ + 0.3$	V
PSRR	Power Supply Rejection Ratio	$V_{DD} = 3\text{V to } 5\text{V}$	70	92		dB
$A_{VOL}$	Open-Loop Large Signal Gain	$V_{OUT} = 0.5\text{V to } 4.5\text{V}$ , $R_{LOAD} = 100\text{k}\Omega$	80	120		dB
$V_{OL}, V_{OH}$	Output Swing from Supply Rail	$R_{LOAD} = 100\text{k}\Omega$		5		mV
$R_{OUT}$	Closed-Loop Output Impedance	$G = 1$ , $f = 1\text{kHz}$ , $I_{OUT} = 0$		0.4		$\Omega$
$R_o$	Open-Loop Output Impedance	$f = 1\text{kHz}$ , $I_{OUT} = 0$		2.6		$\Omega$
$I_{SC}$	Output Short-Circuit Current	Sink or source current		20		mA
$V_{DD}$	Supply Voltage		1.8		6.0	V
$I_q$	Quiescent Current	$V_{DD} = 5\text{V}$		700	950	nA
		$V_{DD} = 2.1\text{V}$		680	900	nA
PM	Phase Margin	$R_{LOAD} = 100\text{k}\Omega$ , $C_{LOAD} = 60\text{pF}$		61		$^\circ$
GM	Gain Margin	$R_{LOAD} = 100\text{k}\Omega$ , $C_{LOAD} = 60\text{pF}$		10		dB
GBWP	Gain-Bandwidth Product	$f = 1\text{kHz}$		18		kHz
$t_s$	Settling Time, 1.5V to 3.5V, Unity Gain	0.1%		0.25		ms
		0.01%		0.253		
		0.1%		0.035		
		0.01%		0.038		
SR	Slew Rate	$A_v = 1$ , $V_{OUT} = 1.5\text{V to } 3.5\text{V}$ , $C_{LOAD} = 60\text{pF}$ , $R_{LOAD} = 100\text{k}\Omega$		10		$\text{mV}/\mu\text{s}$

Typical Performance Characteristics

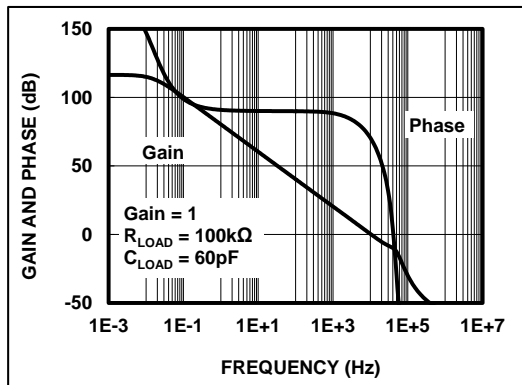
Small-Signal Step Response, 100mV Step



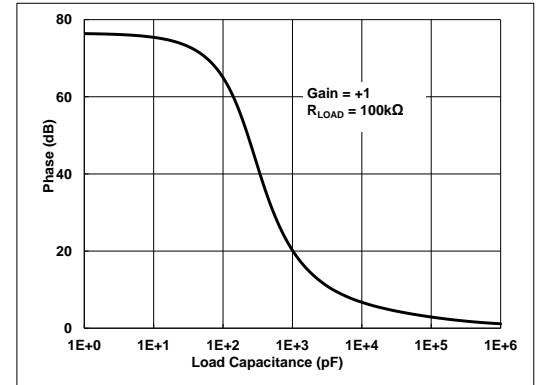
Large-Signal Step Response, 2V Step



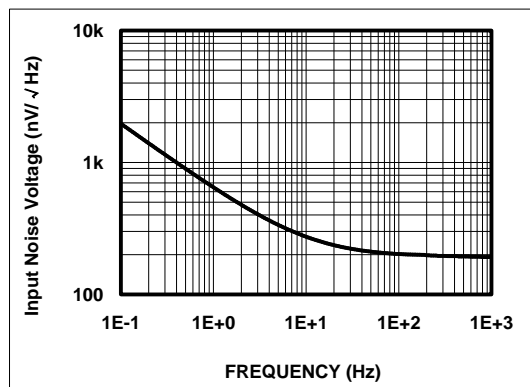
Open-Loop Gain and Phase



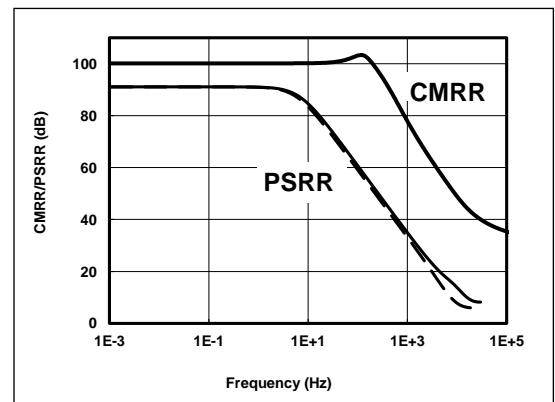
Phase Margin vs.  $C_{LOAD}$  (Stable for Any  $C_{LOAD}$ )



Input Voltage Noise Spectral Density

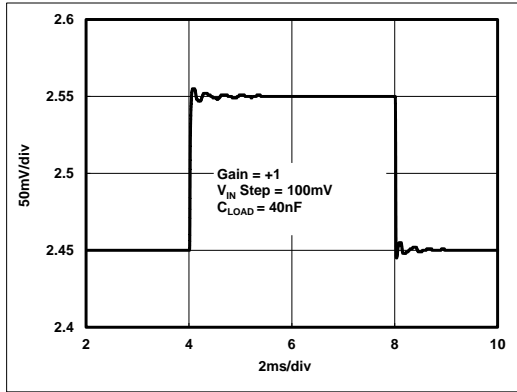


CMRR and PSRR

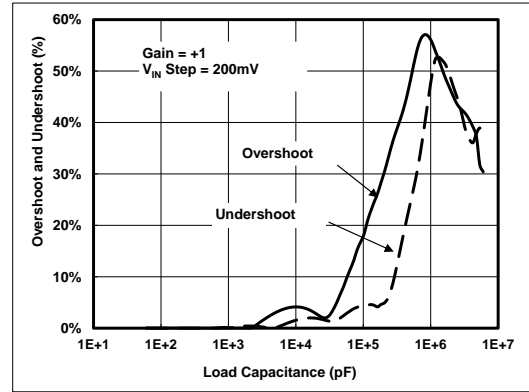


Typical Performance Characteristics

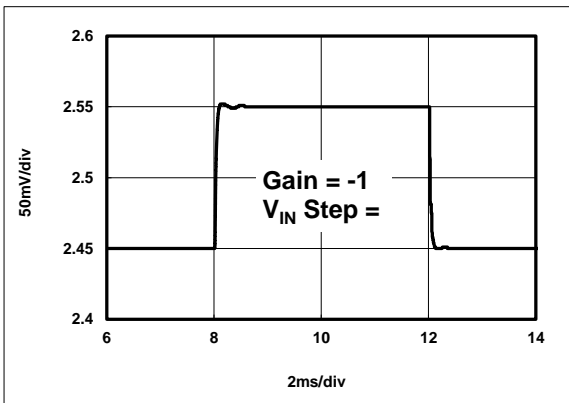
Over-Shoot Voltage,  $C_{LOAD} = 40nF$ , Gain = +1,  $R_{FB}=100k\Omega$



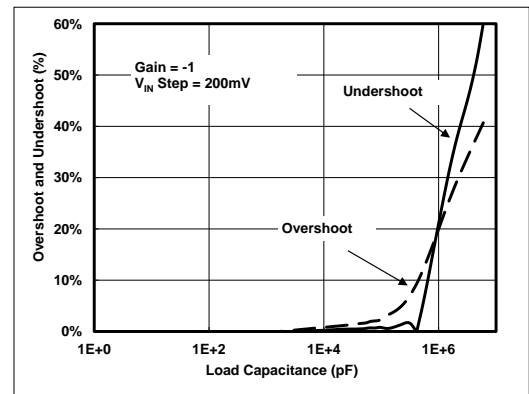
Over-Shoot % vs.  $C_{LOAD}$ , Gain = +1,  $R_{FB} = 1M\Omega$



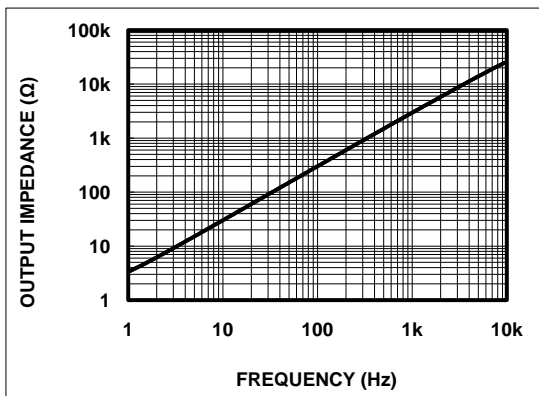
Over-Shoot Voltage,  $C_{LOAD}=40nF$ , Gain= -1,  $R_{FB}=100k\Omega$



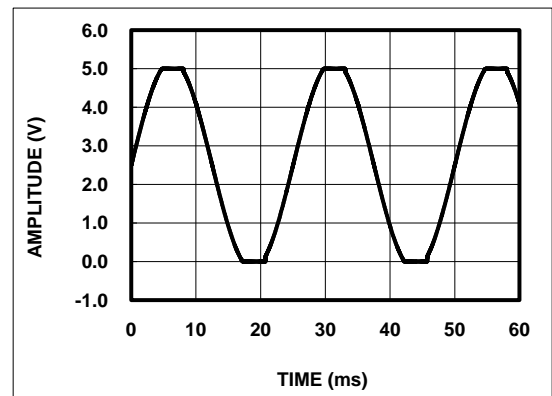
Over-Shoot % vs.  $C_{LOAD}$ , Gain = -1,  $R_{FB} = 1M\Omega$



Closed-Loop Output Impedance vs. Frequency

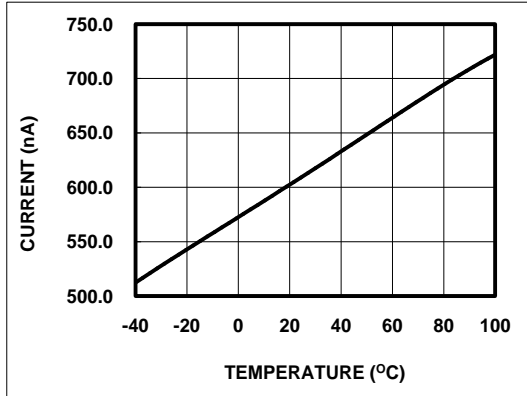


$V_{IN} = -0.2V$  to  $5.7V$ , No Phase Reversal

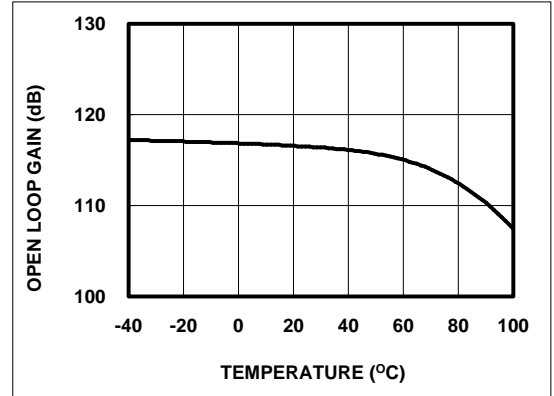


Typical Performance Characteristics

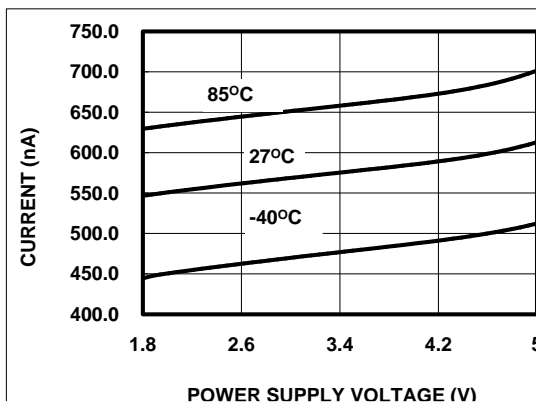
Quiescent Supply Current vs. Temperature



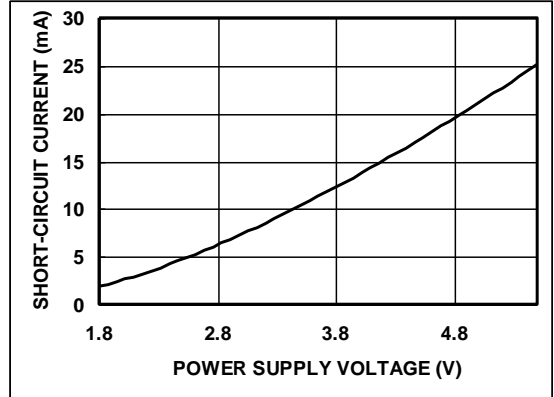
Open-Loop Gain vs. Temperature



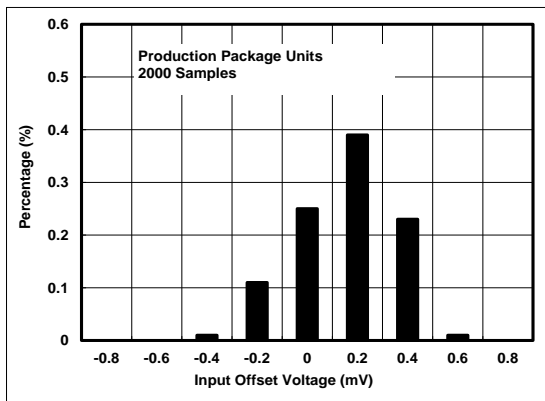
Quiescent Supply Current vs. Supply Voltage



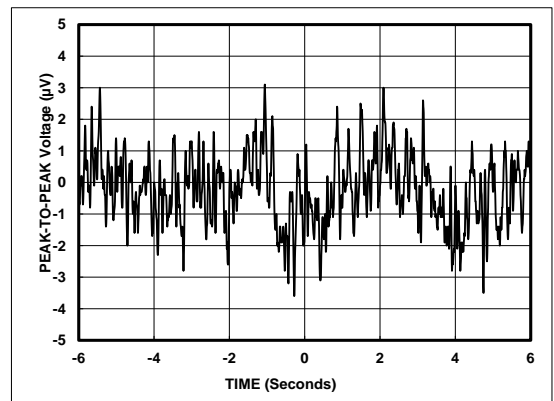
Short-Circuit Current vs. Supply Voltage



Input Offset Voltage Distribution



0.1Hz to 10Hz Time Domain Output Voltage Noise



## Pin Functions

**-IN:** Inverting Input of the Amplifier. Voltage range of this pin can go from  $V^- - 0.3V$  to  $V^+ + 0.3V$ .

**+IN:** Non-Inverting Input of Amplifier. This pin has the same voltage range as -IN.

**V+ or +Vs:** Positive Power Supply. Typically the voltage is from 1.8V to 5.5V. Split supplies are possible as long as the voltage between V+ and V- is between 1.8V and 5.5V. A bypass capacitor of 0.1  $\mu F$  as close to the part as possible should be used between power supply pins or between supply pins and ground.

**V- or -Vs:** Negative Power Supply. It is normally tied to ground. It can also be tied to a voltage other than ground as long as the voltage between V+ and V- is from 1.8V to 5.5V. If it is not connected to ground, bypass it with a capacitor of 0.1  $\mu F$  as close to the part as possible.

**OUT:** Amplifier Output. The voltage range extends to within milli-volts of each supply rail.

## Operation

The TP2191 input signal range extends beyond the negative and positive power supplies. The output can even extend all the way to the negative supply. The input stage is comprised of two CMOS differential amplifiers, a PMOS stage and NMOS stage that are active over different ranges of common mode input voltage. The

Class-AB control buffer and output bias stage uses a proprietary compensation technique to take full advantage of the process technology to drive very high capacitive loads. This is evident from the transient overshoot measurement plots in the Typical Performance Characteristics.

## Applications Information

### Low Supply Voltage and Low Power Consumption

The TP2191 of operational amplifier can operate with power supply voltages from 1.8V to 6.0V. Each amplifier draws only 700 nA quiescent current. The low supply voltage capability and low supply current are ideal for portable applications demanding HIGH CAPACITIVE LOAD DRIVING CAPABILITY and CONSTANT WIDE BANDWIDTH. The TP2191 is optimized for industrial precision and wide bandwidth low power applications. They have an industry leading high GBWP to power ratio and are unity gain stable for 1,000 nF capacitive load. When the load capacitance increases, the increased capacitance at the output pushed the non-dominant pole to lower frequency in the open loop frequency response, lowering the phase and gain margin. Higher gain configurations tend to have better capacitive drive capability than lower gain configurations due to lower closed loop bandwidth and hence higher phase margin.

### Low Input Referred Noise

The TP2191 provides a low input referred noise density of 170 nV/  $\sqrt{Hz}$  at 1 kHz. The voltage noise will grow slowly with the frequency in wideband range, and the input voltage noise is typically 6.5  $\mu V_{P-P}$  at the frequency of 0.1 Hz to 10 Hz.

### Low Input Offset Voltage

The TP2191 has a low offset voltage of 600  $\mu V$  maximum which is essential for precision applications. The offset voltage is trimmed with a proprietary trim algorithm to ensure low offset voltage for precision signal processing requirement.

### Ground Sensing and Rail to Rail Output

The TP2191 has excellent output drive capability, delivering over 10 mA of output drive current. The output stage is a rail-to-rail topology that is capable of swinging to within 5 mV of either rail. Since the inputs can go 300 mV beyond either rail, the op-amp can easily perform 'true ground' sensing.

The maximum output current is a function of total supply voltage. As the supply voltage to the amplifier increases, the output current capability also increases. Attention must be paid to keep the junction temperature of the IC below 150 °C when the output is in continuous short-circuit. The output of the amplifier has reverse-biased ESD diodes connected to each supply. The output should not be forced more than 0.5V beyond either supply, otherwise current will flow through these diodes.

### ESD

The TP2191 has reverse-biased ESD protection diodes on all inputs and output. Input and out pins can not be biased more than 300 mV beyond either supply rail.

### Shut-down

The TP2191N has SHDN pin that can shut down the amplifier to typical 3 nA supply current. The SHDN pin voltage needs to be within 0.5V of V<sub>-</sub> for the amplifier to shut down. During shutdown, the output will be in high output resistance state, which is suitable for multiplexer applications. When left floating, the SHDN pin is internally pulled up to the positive supply and the amplifier remains enabled.

### Driving Large Capacitive Load

The TP2191 op-amp is designed to drive large capacitive loads. Refer to Typical Performance Characteristics for "Phase Margin vs. Load Capacitance". As always, larger load capacitance decreases overall phase margin in a feedback system where internal frequency compensation is utilized. As the load capacitance increases, the feedback loop's phase margin decreases, and the closed-loop bandwidth is reduced. This produces gain peaking in the frequency response, with overshoot and ringing in output step response. The unity-gain buffer (G = +1 V/V) is the most sensitive to large capacitive loads.

When driving large capacitive loads with the TP2191 (e.g., > 200 pF when G = +1 V/V), a small series resistor at the output (R<sub>ISO</sub> in Figure 2) improves the feedback loop's phase margin and stability by making the output load resistive at higher frequencies.

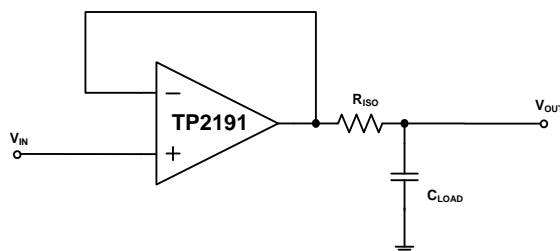


Figure 2

### Power Supply Layout and Bypass

The TP2191 op-amp power supply pin (V<sub>DD</sub> for single-supply) should have a local bypass capacitor (i.e., 0.01 μF to 0.1 μF) within 2 mm for good high frequency performance. It can also use a bulk capacitor (i.e., 1 μF or larger) within 100 mm to provide large, slow currents. This bulk capacitor can be shared with other analog parts.

Ground layout improves performance by decreasing the amount of stray capacitance and noise at the OPA's inputs and outputs. To decrease stray capacitance, minimize PCB board lengths and resistor leads, and place external components as close to the op amps' pins as possible.

### Proper Board Layout

To ensure optimum performance at the PCB level, care must be taken in the design of the board layout. To avoid leakage currents, the surface of the board should be kept clean and free of moisture. Coating the surface creates a barrier to moisture accumulation and helps reduce parasitic resistance on the board.



## Small Size, Precision, 1.8V, 700nA Nanopower, RRIO Op-amps

Keeping supply traces short and properly bypassing the power supplies minimizes power supply disturbances due to output current variation, such as when driving an ac signal into a heavy load. Bypass capacitors should be connected as closely as possible to the device supply pins. Stray capacitances are a concern at the outputs and the inputs of the amplifier. It is recommended that signal traces be kept at least 5 mm from supply lines to minimize coupling.

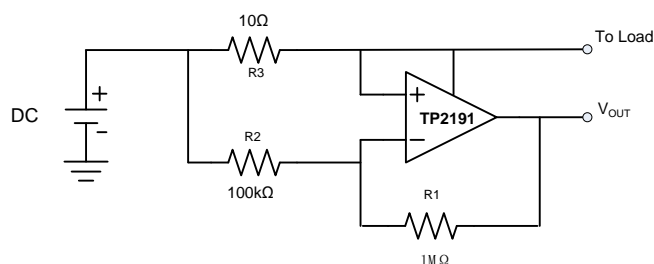
A variation in temperature across the PCB can cause a mismatch in the Seebeck voltages at solder joints and other points where dissimilar metals are in contact, resulting in thermal voltage errors. To minimize these thermocouple effects, orient resistors so heat sources warm both ends equally. Input signal paths should contain matching numbers and types of components, where possible to match the number and type of thermocouple junctions. For example, dummy components such as zero value resistors can be used to match real resistors in the opposite input path. Matching components should be located in close proximity and should be oriented in the same manner. Ensure leads are of equal length so that thermal conduction is in equilibrium. Keep heat sources on the PCB as far away from amplifier input circuitry as is practical.

The use of a ground plane is highly recommended. A ground plane reduces EMI noise and also helps to maintain a constant temperature across the circuit board.

### BATTERY CURRENT SENSING

The Common Mode Input voltage Range of TP2191 op-amp, which goes 0.3V beyond both supply rails, supports their use in high-side and low-side battery current sensing applications. The low quiescent current (700 nA, typical) helps prolong battery life, and the rail-to-rail output supports detection of low currents.

The battery current ( $I_{DD}$ ) through the  $10\ \Omega$  resistor causes its top terminal to be more negative than the bottom terminal. This keeps the Common Mode Input voltage below  $V_{DD}$ , which is within its allowed range. The output of the OPA will also be below  $V_{DD}$ , within its Maximum Output Voltage Swing specification.



$$I_{DD} = \frac{V_{DD} - V_{OUT}}{\frac{R1}{R2} \cdot R3}$$

Figure 3

### Buffered Chemical Sensor (pH) Probe

The TP2191 op-amp has input bias current in the pA range. This is ideal in buffering high impedance chemical sensors such as pH probe. As an example, the circuit in Figure 4 eliminates expensive low-leakage cables that that is required to connect pH probe to metering ICs such as ADC, AFE and/or MCU. A TP2191 op-amp and a lithium battery are housed in the probe assembly. A conventional low-cost coaxial cable can be used to carry OPA's output signal to subsequent ICs for pH reading.

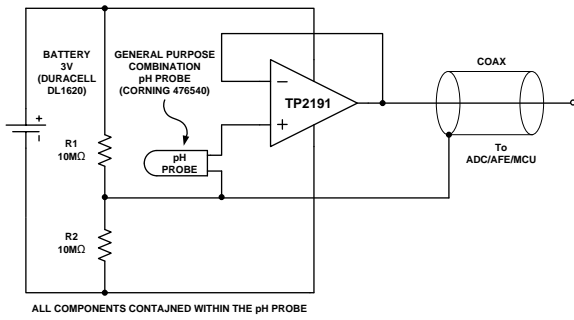


Figure 4: Buffer pH Probe

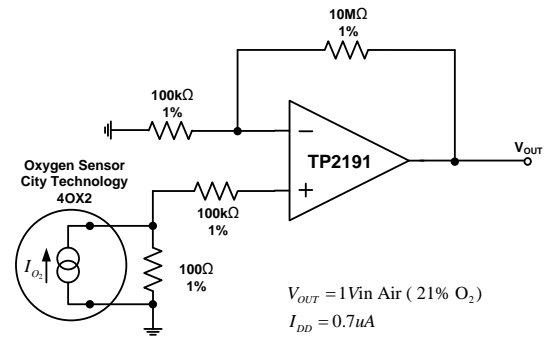


Figure 5

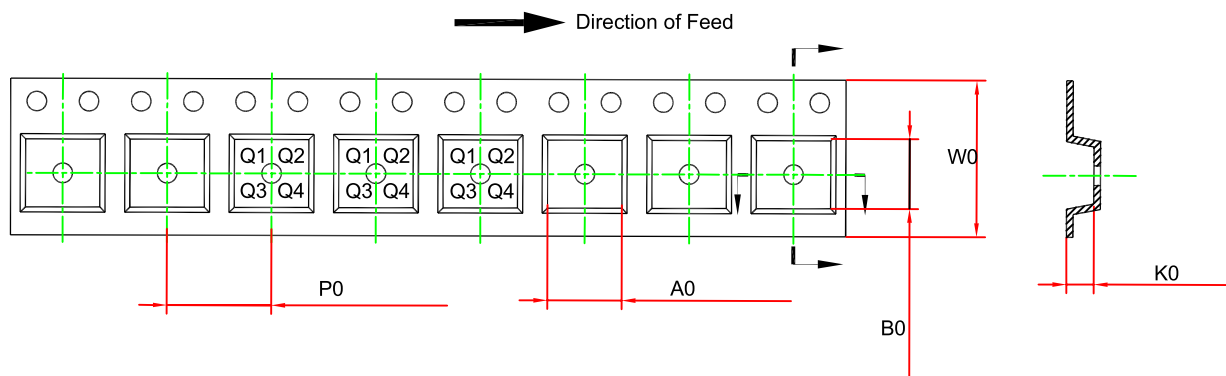
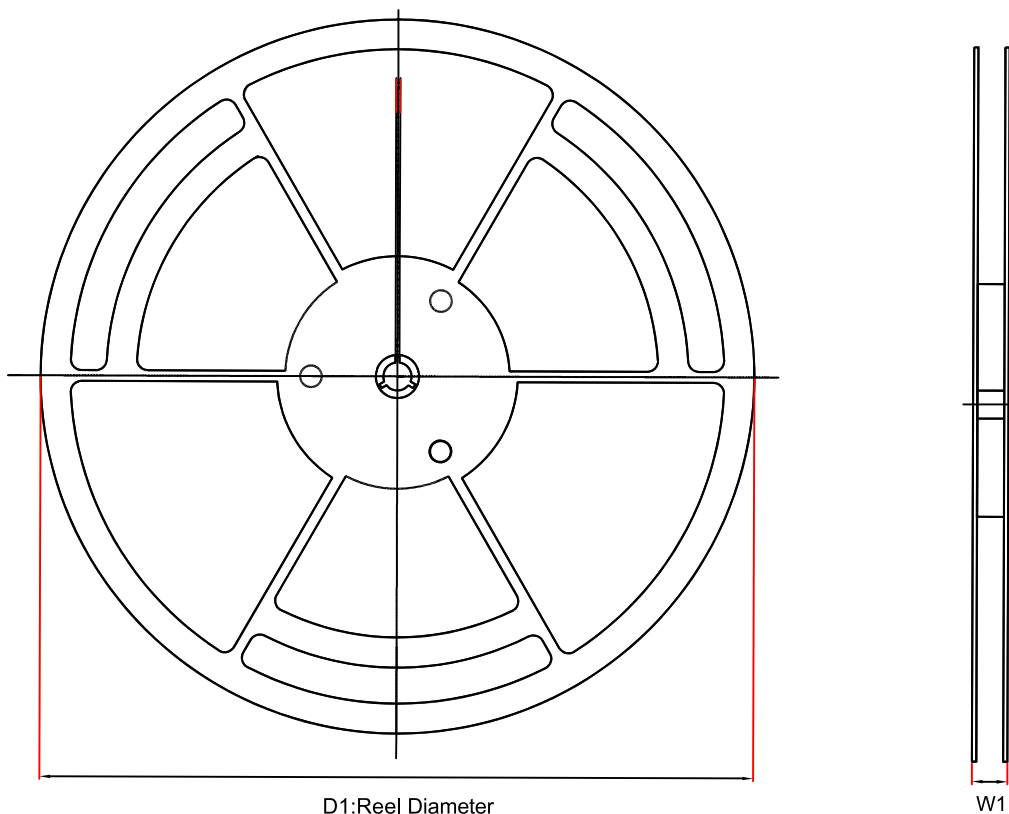
### Portable Gas Sensor Amplifier

Gas sensors are used in many different industrial and medical applications. Gas sensors generate a current that is proportional to the percentage of a particular gas concentration sensed in an air sample. This output current flows through a load resistor and the resultant voltage drop is amplified. Depending on the sensed gas and sensitivity of the sensor, the output current can be in the range of tens of microamperes to a few milli-amperes. Gas sensor datasheets often specify a recommended load resistor value or a range of load resistors from which to choose.

There are two main applications for oxygen sensors – applications which sense oxygen when it is abundantly present (that is, in air or near an oxygen tank) and those which detect traces of oxygen in parts-per-million concentration. In medical applications, oxygen sensors are used when air quality or oxygen delivered to a patient needs to be monitored. In fresh air, the concentration of oxygen is 20.9% and air samples containing less than 18% oxygen are considered dangerous. In industrial applications, oxygen sensors are used to detect the absence of oxygen; for example, vacuum-packaging of food products.

The circuit in Figure 5 illustrates a typical implementation used to amplify the output of an oxygen detector. With the components shown in the figure, the circuit consumes less than 700 nA of supply current ensuring that small form-factor single- or button-cell batteries (exhibiting low mAh charge ratings) could last beyond the operating life of the oxygen sensor. The precision specifications of this amplifier, such as their low offset voltage, low  $V_{OS}$  TC, low input bias current, high CMRR, and high PSRR are other factors which make these amplifiers excellent choices for this application.

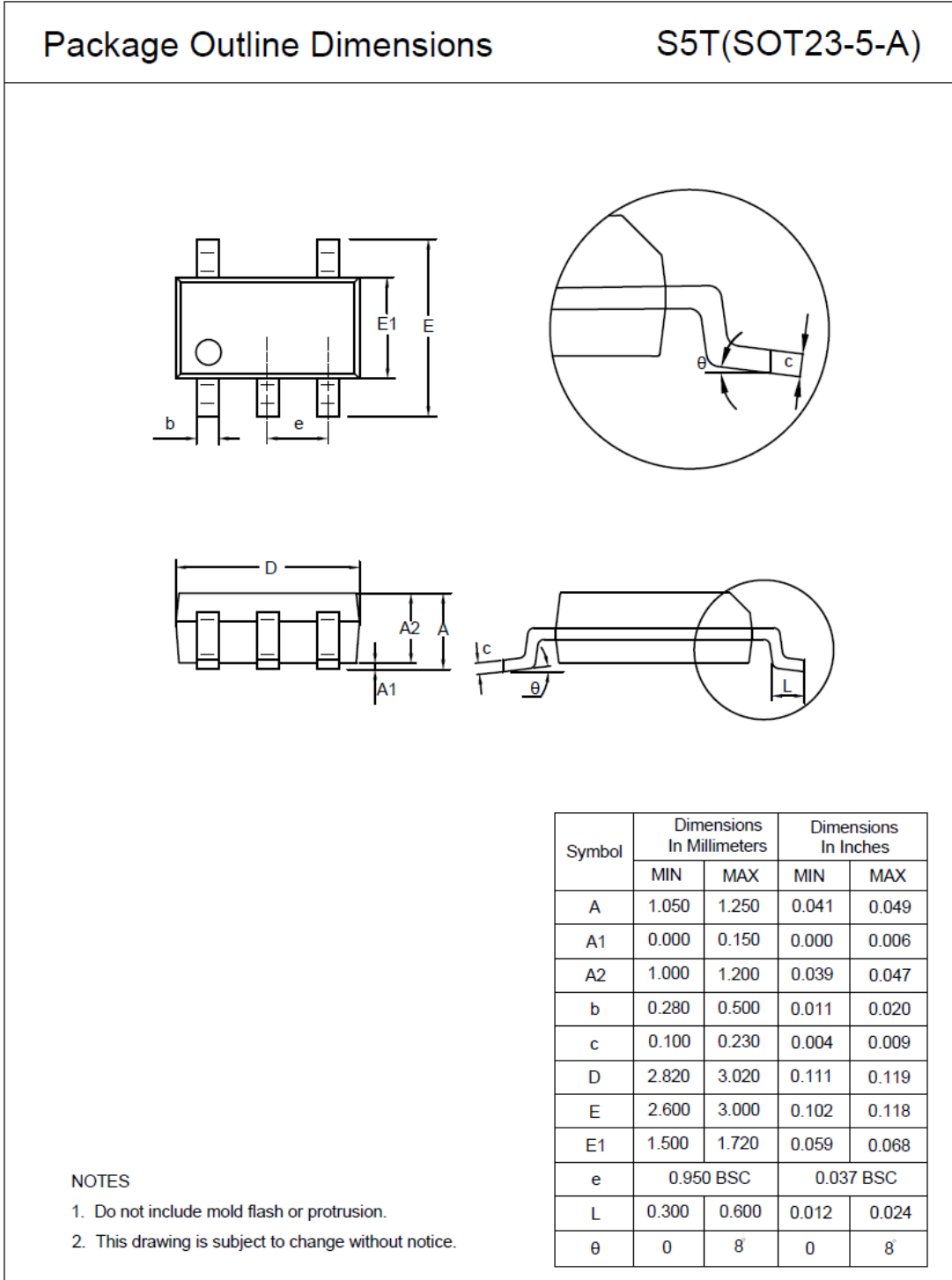
### Tape and Reel Information



Order Number	Package	D1 (mm)	W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	W0 (mm)	Pin1 Quadrant
TP2191-TR	SOT23-5	178.0	12	3.3	3.25	1.4	4.0	8.0	Q3

Package Outline Dimensions

SOT23-5



## Revision History

Date	Revision	Notes
2023-03-15	Rev.1.2	The following updates are all about the new datasheet formats or typo, the actual product remains unchanged. Updated marking information of TP2191-TR: B2T to B2X. Updated to new format of package dimensions. Updated ESD value with new test standard. Updated specification with test limit. Added MSL information. Added tape and reel information.